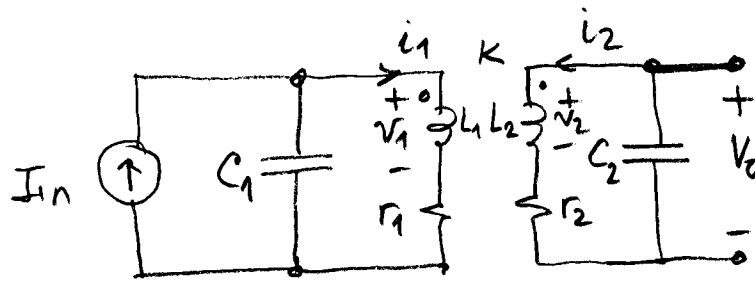


(2)



$$\omega_0^2 = \frac{1}{L_1 C_1} = \frac{1}{L_2 C_2}$$

(1)

Odrediti koeficijent sprege kalimova K tako da f -ja prenosa $R(s) = \frac{V_0(s)}{I_{in}(s)}$ bude sa

- Butterworthovom karakteristikom (maksimalno ravno)
- Chebyshevovom karakteristikom (equi-ripple) kod koje je $r = 0.5 \text{ dB}$ (ripple)

(P)

$$v_1(t) = L_1 \frac{di_1(t)}{dt} + M \frac{di_2(t)}{dt}$$

$$v_2(t) = L_2 \frac{di_2(t)}{dt} + M \frac{di_1(t)}{dt}$$

$$M = K \sqrt{L_1 L_2}$$

$$n = \sqrt{\frac{L_1}{L_2}}$$

$$V_1(s) = sL_1 I_1(s) + sM I_2(s)$$

$$V_2(s) = sL_2 I_2(s) + sM I_1(s)$$

KZ:

$$I_{in}(s) = I_1(s) + sC_1 (V_1(s) + r_1 I_1(s))$$

$$V_0(s) = V_2(s) + r_2 I_2(s), \quad sC_2 V_0(s) = -I_2(s)$$

$$\Rightarrow V_0(s) = \frac{V_2(s)}{1 + s r_2 C_2}$$

$$\Rightarrow \frac{V_0(s)}{I_{in}(s)} = \frac{sM}{(1 + s r_1 C_1 + s^2 L_1 C_1)(1 + s r_2 C_2 + s^2 L_2 C_2) - s^4 M^2 C_1 C_2}$$

$$\omega_0^2 = \frac{1}{L_1 C_1} = \frac{1}{L_2 C_2}, \quad Q_1 = \frac{1}{\omega_0 L_1 r_1}, \quad Q_2 = \frac{1}{\omega_0 L_2 r_2}$$

$$\Rightarrow R(s) = \frac{V_0(s)}{I_{in}(s)} = \frac{sM \omega_0^4 Q_1 Q_2}{(s^2 Q_1 + s\omega_0 + \omega_0^2 Q_1)(s^2 Q_2 + s\omega_0 + \omega_0^2 Q_2) - s^4 K^2 Q_1 Q_2}$$

polovi:

$$s^4 K^2 Q_1 Q_2 - (s^2 Q_1 + s\omega_0 + \omega_0^2 Q_1)(s^2 Q_2 + s\omega_0 + \omega_0^2 Q_2) = 0$$

$$\Rightarrow [s^2 Q_1(1-K) + s\omega_0 + \omega_0^2 Q_1][s^2 Q_2(1+K) + s\omega_0 + \omega_0^2 Q_2] = 0$$

$$s_{1,2} = -\frac{\omega_0}{2Q_1(1-K)} \pm j \frac{\omega_0}{Q_1(1-K)} \sqrt{4Q_1^2(1-K) - 1}$$

$$S_{3,4} = -\frac{\omega_0}{2Q_2(1+K)} \mp j \frac{\omega_0}{Q_2(1+K)} \sqrt{4Q_2^2(1+K)-1}$$

$$4Q_1^2(1-K) \gg 1, 4Q_2^2(1+K) \gg 1$$

$$\Rightarrow S_{1,2} \approx -\frac{\omega_0}{2Q_1(1-K)} \mp j \frac{\omega_0}{\sqrt{1-K}}$$

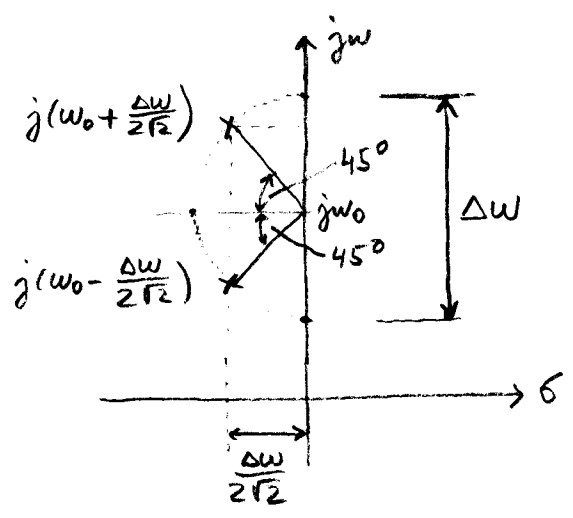
$$S_{3,4} \approx -\frac{\omega_0}{2Q_2(1+K)} \mp j \frac{\omega_0}{\sqrt{1+K}}$$

kada je koeficijent sprege $K \ll 1$

$$S_{1,2} \approx -\frac{\omega_0}{2Q_1(1-K)} \mp j\omega_0(1+\frac{K}{2}), (1+K)^{-\frac{1}{2}} \approx 1 - \frac{1}{2}K$$

$$S_{3,4} \approx -\frac{\omega_0}{2Q_2(1+K)} \mp j\omega_0(1-\frac{K}{2}), (1-K)^{-\frac{1}{2}} \approx 1 + \frac{1}{2}K$$

a) položaj polova kod Butterworthovog band-pass filtra



$$R(S_{1,2}) = -\frac{\omega_0}{2Q_1(1-K)} \approx -\frac{\omega_0}{2Q_1}$$

$$R(S_{3,4}) = -\frac{\omega_0}{2Q_2(1+K)} \approx -\frac{\omega_0}{2Q_2}$$

$$\Rightarrow Q_1 \approx Q_2 = Q$$

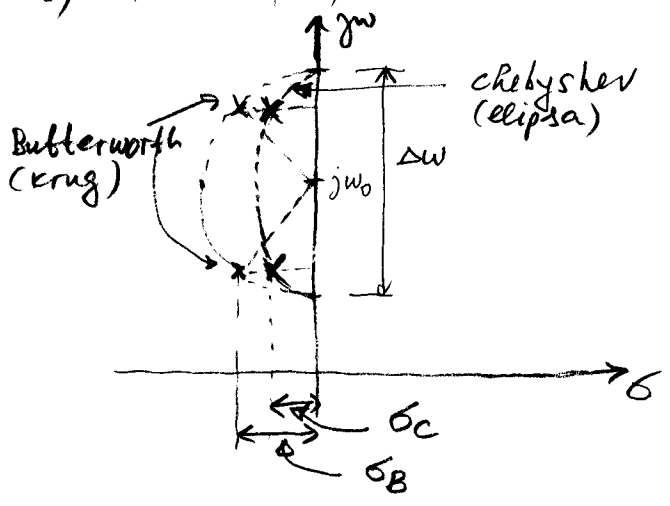
$$\frac{\Delta \omega}{2\sqrt{2}} = \frac{\omega_0}{2Q} \Rightarrow \Delta \omega = \sqrt{2} \frac{\omega_0}{Q}$$

$$\omega_0(1 + \frac{K}{2}) = \omega_0(1 + \frac{\Delta \omega}{2\sqrt{2}\omega_0})$$

$$\frac{K}{2} = \frac{\Delta \omega}{2\sqrt{2}\omega_0} = \frac{\omega_0}{2Q \cdot \omega_0}$$

$$\Rightarrow K = \frac{1}{Q} \Rightarrow B = \sqrt{2} \frac{f_0}{Q} !$$

b) položaj polova kod chebyshevove K-K

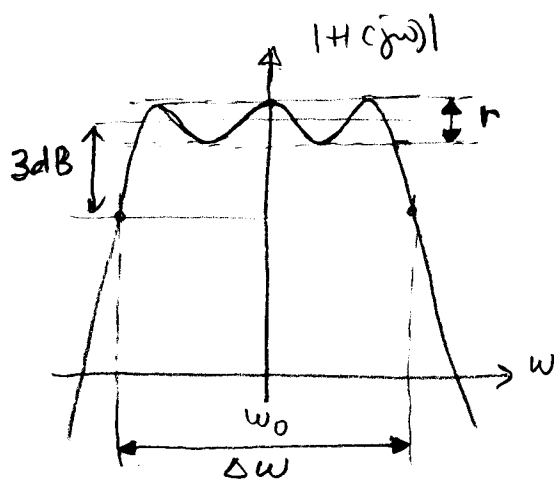


$$(\sigma_i)_c = (\tanh a) (\sigma_i)_B$$

$$a = \frac{1}{n} \operatorname{asinh} \frac{1}{\sqrt{\epsilon}} = 10^{\frac{r_{dB}}{10}} - 1$$

r-pipple

r _{dB}	n=2	n=3	n=4
0.05	0.898	0.750	0.623
0.1	0.859	0.696	0.557
0.2	0.806	0.631	0.505
0.3	0.767	0.588	0.467
0.4	0.736	0.556	0.439
0.5	0.709	0.524	0.416



uzmimo $r = 0.5 \text{ dB}$

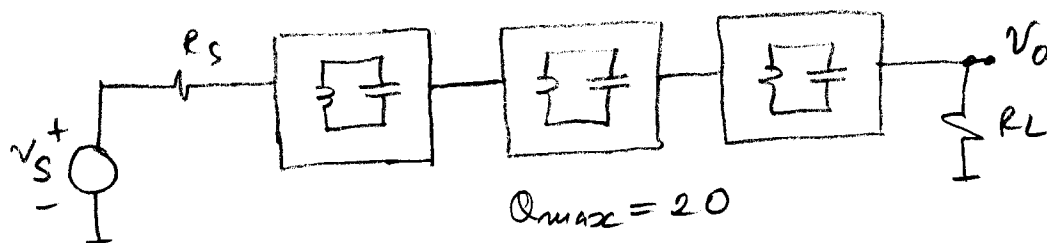
$$\Rightarrow \sigma_c = \frac{\omega_0}{2Q} = 0.709 \omega_0 \frac{K}{2}$$

$$\Rightarrow K = \frac{1}{0.709} \frac{1}{Q} \approx \frac{\sqrt{2}}{Q}$$

$$\Delta\omega = 2\sqrt{2} \omega_0 \frac{K}{2} = 2 \frac{\omega_0}{Q}$$

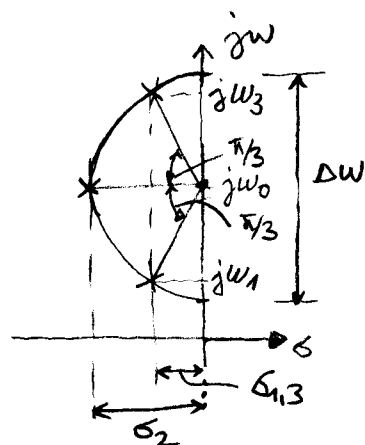
$$B = 2 \frac{f_0}{Q} \text{ ?}$$

2



Tri selektivne pojačavača vezane su kasadno radi dobijanja veceg pojačanja i maksimalno ravne frekventne karakteristike. Ako je centralna učestavost $f_0 = 16 \text{ MHz}$, odrediti položaj polova selektivnih pojačavača.

R



$$\sigma_2 = \frac{\omega_0}{2Q_2} \quad \sigma_{1,3} = \frac{\omega_1}{2Q_1} = \frac{\omega_3}{2Q_3}$$

$$\sigma_2 = \frac{\Delta\omega}{2} \Rightarrow Q_2 = \frac{\omega_0}{\Delta\omega} = \frac{f_0}{\Delta f}$$

$$\sigma_1 = \frac{\omega_1}{2Q_1} = \frac{\Delta\omega}{4} \Rightarrow Q_1 = \frac{2\omega_1}{\Delta\omega} = \frac{2f_1}{\Delta f}$$

$$\sigma_3 = \frac{\omega_3}{2Q_3} = \frac{\Delta\omega}{4} \Rightarrow Q_3 = \frac{2\omega_3}{\Delta\omega} = \frac{2f_3}{\Delta f}$$

Slika: $\omega_1 = \omega_0 - \frac{\Delta\omega}{2} \cos \frac{\pi}{6}$
 $\omega_3 = \omega_0 + \frac{\Delta\omega}{2} \cos \frac{\pi}{6}$

maksimalni Q faktor treba da ima treći pojačavač

$$Q_3 = Q_{max} = 20 = \frac{2f_3}{\Delta f} = \frac{2f_0}{\Delta f} + \cos \frac{\pi}{6}$$

$$\Rightarrow \Delta f = 104.5 \text{ MHz}$$

$$f_1 = f_0 - \frac{\Delta f}{2} \cos \frac{\pi}{6} = 954.74 \text{ MHz} \Rightarrow Q_1 = \frac{2f_1}{\Delta f} = 18.3$$

$$Q_2 = \frac{f_0}{\Delta f} = 9.6$$

$$f_3 = f_0 + \frac{\Delta f}{2} \cos \frac{\pi}{6} = 1045.26 \text{ MHz}$$